Supplemental Materials to “A PID Controller Approach for Stochastic Optimization of Deep Networks”

1. Connection

In this section, we show the mathematical connection between deep learning optimizer and PID controller.

1.1. PID Controller

The PID controller is the most commonly used control algorithm in industry since its origin in the 1940s. More than 90% of the controllers in the industrial products are PID [2]. Which has the following definition:

\[ u(t) = K_p e(t) + K_i \sum_{i=0}^{t-1} e(i) + K_d (e(t) - e(t-1)) \]  

(1)

where \( u(t) \) is the controller’s update, and \( e(t) \) is the error between the system’s output and the desired output. \( K_p \), \( K_i \) and \( K_d \) are positive constants to balance present, past and future of the error \( e(t) \).

By replacing the error \( e(t) \) in PID controller with the gradient in deep learning optimization, the PID controller for deep learning optimization is given by:

\[ u(t) = K_p \frac{\partial L_t}{\partial \theta_t} + K_i \sum_{i=0}^{t-1} \left( \frac{\partial L_t}{\partial \theta_i} \right) + K_d \left( \frac{\partial L_t}{\partial \theta_t} - \frac{\partial L_{t-1}}{\partial \theta_{t-1}} \right) \]  

(2)

where \( u(t) \) is the update of the weight, \( \theta_t \) is the weight at iteration \( t \) and \( \frac{\partial L_t}{\partial \theta_t} \) is the gradient of the network.

1.2. Deep Learning Optimizers

1.2.1 SGD is a P Controller

The update rule of SGD is:

\[ \theta_{t+1} - \theta_t = -r \frac{\partial L_t}{\partial \theta_t} \]  

(3)

where \( r \) is the learning rate.

Comparing Equation 3 with Equation 2, we can see that the update of parameters relies on current gradient, and SGD is a P controller.

1.2.2 SGD-Momentum is a PI Controller

The update rule of SGD-Momentum is given by:

\[
\begin{aligned}
V_{t+1} &= \alpha V_t - r \frac{\partial L_t}{\partial \theta_t} \\
\theta_{t+1} &= \theta_t + V_{t+1}
\end{aligned}
\]  

(4)

where \( \alpha \) is a value to balance past and current gradients, usually set to 0.9 [3]. Dividing both sides of the 1st formula of Equation 4 by \( \alpha^{t+1} \):
\[
\frac{V_{t+1}}{\alpha^{t+1}} = \frac{V_t}{\alpha^t} - r \frac{\partial L_t}{\partial \theta_t} \frac{1}{\alpha^{t+1}}
\]  

(5)

By applying Equation 5 from time \(t + 1\) to 1, we have:

\[
\begin{align*}
\frac{V_{t+1}}{\alpha^{t+1}} - \frac{V_t}{\alpha^t} &= -r \frac{\partial L_t}{\partial \theta_t} \\
\frac{V_t}{\alpha^t} - \frac{V_{t-1}}{\alpha^{t-1}} &= -r \frac{\partial L_{t-1}}{\partial \theta_{t-1}} \\
&\quad \vdots \\
\frac{V_1}{\alpha^1} - \frac{V_0}{\alpha^0} &= -r \frac{\partial L_0}{\partial \theta_0}
\end{align*}
\]  

(6)

Add the above \(t + 1\) equations together, there is:

\[
\frac{V_{t+1}}{\alpha^{t+1}} = \frac{V_0}{\alpha^0} - r \left( \sum_{i=0}^{t} \left( \frac{\partial L_i}{\alpha^{t+1}} \right) \right)
\]  

(7)

Without loss of generality, we set the initial condition \(V_0 = 0\), and thus the above equation can be simplified as follows:

\[
V_{t+1} = -r \left( \sum_{i=0}^{t} (\alpha^{t-i} \partial L_i / \partial \theta_{t-i}) \right)
\]  

(8)

Put \(V_{t+1}\) into the 2nd formula of Equation 4, we have:

\[
\theta_{t+1} - \theta_t = -r \frac{\partial L_t}{\partial \theta_t} - r \left( \sum_{i=0}^{t-1} (\alpha^{t-i} \partial L_i / \partial \theta_i) \right)
\]  

(9)

We can see that the update of the parameter relies on both the current gradient (P control) and the integral of past gradients (I control). If we assume \(\alpha = 1\), there is:

\[
\theta_{t+1} - \theta_t = -r \left( \partial L_t / \partial \theta_t \right) - r \left( \sum_{i=0}^{t-1} (\partial L_i / \partial \theta_i) \right)
\]  

(10)

Comparing Equation 10 with Equation 2, we can see that SGD-Momentum is a PI controller with \(K_p = r\) and \(K_i = r\).

### 1.2.3 Nesterov’s Momentum is a PI Controller with larger P

The update rule of Nesterov’s Momentum is:

\[
\begin{align*}
V_{t+1} &= \alpha V_t - r \partial L_t / \partial (\theta_t + \alpha V_t) \\
\hat{\theta}_{t+1} &= \hat{\theta}_t + V_{t+1}
\end{align*}
\]  

(11)

The expression is almost the same as SGD-Momentum except for the location where the gradient is evaluated. By using a variable transform \(\hat{\theta}_t = \theta_t + \alpha \ast V_t\), we have:

\[
\begin{align*}
V_{t+1} &= \alpha V_t - r \partial L_t / \partial \hat{\theta}_t \\
\hat{\theta}_{t+1} &= \hat{\theta}_t + (1 + \alpha) V_{t+1} - \alpha V_t
\end{align*}
\]  

(12)

Similar to the derivation process in Equations 5, 6 and 7 of SGD-Momentum, we have:

\[
V_{t+1} = -r \left( \sum_{i=1}^{t} (\alpha^{t-i} \partial L_i / \partial \hat{\theta}_i) \right)
\]  

(13)
With Equation. 13, Equation. 11 can be rewritten as:

$$\hat{\theta}_{t+1} - \hat{\theta}_t = -r(1 + \alpha)\partial L_t / \partial \hat{\theta}_t - \alpha r \sum_{i=1}^{t-1} \frac{\partial L_i / \partial \hat{\theta}_i}{\partial \hat{\theta}_i}$$  \hspace{1cm} (14)$$

One can see that the update of parameters relies on the current gradient (P control) and the integral of past gradients (I control). If we assume $\alpha = 1$, then:

$$\hat{\theta}_{t+1} - \hat{\theta}_t = -2r(\partial L_t / \partial \hat{\theta}_t) - r \left( \sum_{i=0}^{t-1} \frac{\partial L_i / \partial \hat{\theta}_i}{\partial \hat{\theta}_i} \right)$$  \hspace{1cm} (15)$$

Comparing Equation. 15 with Equation. 2, we can see that Nesterov's Momentum is a PI controller with $K_p = 2r$ and $K_i = r$.

2. Laplace Transform of PID Optimizer

2.1. Laplace Transform

The Laplace Transform converts the function of real variable $t$ (iteration) to a function of complex variable $s$ (frequency). Denote by $F(s)$ the Laplace transform of $f(t)$. There is

$$F(s) = \int_0^{\infty} e^{-st} f(t) \, dt, \text{ for } s > 0.$$  \hspace{1cm} (16)$$

Usually $F(s)$ is easier to solve than $f(t)$, and $f(t)$ can be recovered from $F(s)$ by the Inverse Laplace Transform:

$$f(t) = \frac{1}{2\pi i} \lim_{\gamma \to \infty} \int_{\gamma - iT}^{\gamma + iT} e^{st} F(s) \, ds,$$  \hspace{1cm} (17)$$

where $\gamma$ is a real number and $i$ is the unit of imagery part. In practice, we could decompose a Laplace transform into known transforms of functions in the Laplace table [5], which includes most of the commonly used Laplace transforms, and then construct the inverse transform.

With Laplace Transform, we convert the PID optimizer into its Laplace transformed functions of $s$, and then simplify the algebra. Once we find the transformed solution of $F(s)$, we can inverse the transform to obtain the required solution $f$ as a function of $t$.

2.2. Evolution of Weight

A weight of a deep model is initialized as a scalar $\theta_0$, and it is updated iteratively to reach its optimal value, denoted by $\theta_\ast$. Then the process of each weight in DNN can be viewed as a step response (from $\theta_0$ to $\theta_\ast$) in control theory [4]. We then use the Laplace Transform as a guide to set hyper-parameter $K_d$.

The Laplace Transform of $\theta_\ast$ is $\frac{\theta_\ast}{s}$ [5]. We denote by $\theta(t)$ the weight at iteration $t$. The Laplace Transform of $\theta(t)$ is denoted as $\theta(s)$, and that of error $e(t)$ as $E(s)$. Since $E(s) = \frac{\theta_\ast}{s} - theta(s)$. The Laplace transform of PID [5] is:

$$U(s) = (K_p + K_i \frac{1}{s} + K_d s) E(s)$$  \hspace{1cm} (18)$$

In our case, the $u(t)$ corresponds to the update of $\theta(t)$. So we replace $U(s)$ with $\theta(s)$, and with $E(s) = \frac{\theta_\ast}{s} - \theta(s)$, Equation. 18 can be rewritten as:

$$\theta(s) = (K_p + K_i \frac{1}{s} + K_d s) \left( \frac{u_s}{s} - \theta(s) \right)$$  \hspace{1cm} (19)$$

With this form, it is easy to derive a standard closed loop transfer function [1] as follows:

$$\frac{\theta_\ast}{s} - \theta(s) = \frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$  \hspace{1cm} (20)$$
where

\[
\begin{align*}
\frac{K_p+1}{K_d} &= 2\zeta \omega_n \\
\frac{K_p}{K_d} &= \omega_n
\end{align*}
\] (21)

Equation 20 can be rewritten as:

\[
\frac{\theta_* - \theta(s)}{s} = \frac{(s + \zeta \omega_n) + \frac{\zeta}{\sqrt{1+\zeta^2}} \omega_n \sqrt{1-\zeta^2}}{(s + \zeta \omega_n)^2 + \omega_n^2(1-\zeta^2)}
\] (22)

We can get the time (iteration) domain form of \( \theta(s) \) by using the Laplace Inverse Transform table [5] and the initial condition of the \( \theta(\theta_0) \):

\[
\theta(t) = \theta_* - \frac{(\theta_* - \theta_0)\sin(\omega_n \sqrt{1-\zeta^2} t + \arccos(\zeta))}{e^{\zeta \omega_n t} \sqrt{1-\zeta^2}}
\] (23)

where \( \zeta \) and \( \omega_n \) are damping ratio and natural frequency of the system, respectively. In Fig. 1, we show the evolution process of a weight as an example of \( \theta(t) \). From Equation (21), we can write \( \zeta = \frac{(K_p+1)^2}{4K_dK_i} \). One can see that \( K_i \) is a monotonically decreasing function of \( \zeta \). Refer to the definition of overshoot:

\[
\text{Overshoot} = \frac{\theta_{max} - \theta_*}{\theta_*}
\] (24)

By differentiating \( \theta(t) \) w.r.t. time \( t \), and let \( d\theta(t)/dt = 0 \), we have the peak time of the weight as:

\[
t_{max} = \frac{\pi}{\omega_n \sqrt{1-\zeta^2}}
\] (25)

Put \( t_{max} \) to Equation 23, we have \( \theta_{max} \), and put \( \theta_{max} \) to Equation 24, we have:

\[
\text{Overshoot} = \frac{\theta(t_{max}) - \theta_*}{\theta_*} = e^{-\frac{\pi \zeta}{\omega_n \sqrt{1-\zeta^2}}}
\] (26)

One can see that \( \zeta \) is monotonically decreasing with overshoot. Then \( K_i \) is a monotonically increasing function of overshoot. So more history error (Integral part), more overshoot the system will have. That is the reason why SGD-Momentum which accumulates past gradients will overshoot its target and spend more time during training.

As can be observed from Equation. (23), the term \( \sin(\omega_n \sqrt{1-\zeta^2} t + \arccos(\zeta)) \) brings periodically oscillation change to the weight, which is no more than 1. The term \( e^{-\zeta \omega_n t} \) mainly controls the convergence rate. There is a hyper-parameter \( K_d \) in calculating the derivate \( e^{-\zeta \omega_n} = e^{-\frac{K_p+1}{2K_d}} \). It is easy to observe that the larger the derivate, the earlier the training convergence we will reach. However, when \( K_d \) gets too large, the system will be fragile. In practice, we set the hyper-parameter \( K_d \) based...
on the Ziegler-Nichols optimum setting rule [6], which is widely used by engineers in PID feedback control since its origin in 1940s.

According to Ziegler-Nichols’ rule, the ideal setup of $K_d$ should be one third of the oscillation period, which means $K_d = \frac{1}{3}T$, where $T$ is the period of oscillation. From Equation. (23), we can get $T = \frac{2\pi n}{\omega_n \sqrt{1 - \zeta^2}}$. If we make a simplification that $\alpha$ in Momentum is equal to 1, then $K_i = K_d = r$. Combined with Equation. (21), $K_d$ will have a closed form solution:

$$K_d = 0.25r + 0.5 + \frac{16}{9} \pi^2 / r$$

(27)

References